

# Channel capacity estimation using free probability theory

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January 2008

# Problem at hand

The capacity per receiving antenna of a channel with  $n \times m$  channel matrix  $\mathbf{H}$  and signal to noise ratio  $\rho = \frac{1}{\sigma^2}$  is given by

$$C = \frac{1}{n} \log_2 \det \left( \mathbf{I}_n + \frac{1}{m\sigma^2} \mathbf{H}\mathbf{H}^H \right) = \frac{1}{n} \sum_{l=1}^n \log_2 \left( 1 + \frac{1}{\sigma^2} \lambda_l \right) \quad (1)$$

where  $\lambda_l$  are the eigenvalues of  $\frac{1}{m} \mathbf{H}\mathbf{H}^H$ . We would like to estimate  $C$ .

To estimate  $C$ , we will use free probability tools to estimate the eigenvalues of  $\frac{1}{m} \mathbf{H}\mathbf{H}^H$  based on some observations  $\hat{\mathbf{H}}$ ;

# Observation model 1

The following is a much used observation model:

$$\hat{\mathbf{H}}_i = \mathbf{H} + \sigma \mathbf{X}_i \quad (2)$$

where

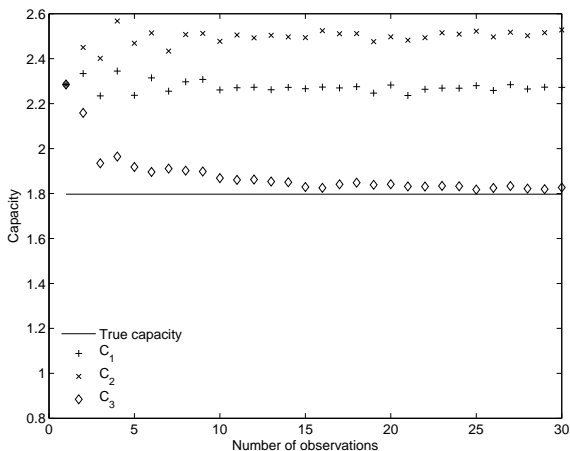
- ▶ The matrices are  $n \times m$  ( $n$  is the number of receiving antennas,  $m$  is the number of transmitting antennas)
- ▶  $\hat{\mathbf{H}}_i$  is the measured MIMO matrix,
- ▶  $\mathbf{X}_i$  is the noise matrix with i.i.d standard complex Gaussian entries.

# Existing ways to estimate the channel capacity

Several channel capacity estimators have been used in the literature:

$$\begin{aligned}
 C_1 &= \frac{1}{nL} \sum_{i=1}^L \log_2 \det \left( \mathbf{I}_n + \frac{1}{m\sigma^2} \hat{\mathbf{H}}_i \hat{\mathbf{H}}_i^H \right) \\
 C_2 &= \frac{1}{n} \log_2 \det \left( \mathbf{I}_n + \frac{1}{L\sigma^2 m} \sum_{i=1}^L \hat{\mathbf{H}}_i \hat{\mathbf{H}}_i^H \right) \\
 C_3 &= \frac{1}{n} \log_2 \det \left( \mathbf{I}_n + \frac{1}{\sigma^2 m} \left( \frac{1}{L} \sum_{i=1}^L \hat{\mathbf{H}}_i \right) \left( \frac{1}{L} \sum_{i=1}^L \hat{\mathbf{H}}_i \right)^H \right)
 \end{aligned} \tag{3}$$

Why not try to formulate an estimator based on free probability instead?



Comparison of the classical capacity estimators for various number of observations.  $\sigma^2 = 0.1$ ,  $n = 10$  receive antennas,  $m = 10$  transmit antennas. The rank of  $\mathbf{H}$  was 3.

# The Marčenko Pastur law

The Marčenko Pastur law  $\mu_c$ :

$$f^{\mu_c}(x) = \left(1 - \frac{1}{c}\right)^+ \delta_0(x) + \frac{\sqrt{(x-a)^+(b-x)^+}}{2\pi cx}, \quad (4)$$

where  $(z)^+ = \max(0, z)$ ,  $a = (1 - \sqrt{c})^2$ ,  $b = (1 + \sqrt{c})^2$ , and  $\delta_0(x)$  is dirac measure (point mass) at 0.

- ▶ free cumulants:  $1, c, c^2, c^3, \dots$
- ▶  $\mu_c$  is the limit eigenvalue distribution of  $\frac{1}{N} \mathbf{X} \mathbf{X}^H$ , with  $\mathbf{X}$  an  $n \times N$  with independent standard complex Gaussian entries as  $N \rightarrow \infty$ , and  $\frac{n}{N} \rightarrow c$ .

# Main free probability result we will use

Define

$$\begin{aligned}\Gamma_n &= \frac{1}{N} \mathbf{R}_n \mathbf{R}_n^H \\ \mathbf{W}_n &= \frac{1}{N} (\mathbf{R}_n + \sigma \mathbf{X}_n) (\mathbf{R}_n + \sigma \mathbf{X}_n)^H,\end{aligned}$$

where  $\mathbf{R}_n$  and  $\mathbf{X}_n$  are independent  $n \times N$  random matrices,  $\mathbf{X}_n$  is complex, standard, Gaussian.

## Theorem

If e.e.d.  $(\Gamma_n) \rightarrow \nu_\Gamma$ , then e.e.d.  $(\mathbf{W}_n) \rightarrow \nu_W$  where  $\nu_W$  is uniquely identified by

$$\nu_W \boxminus \mu_c = (\nu_\Gamma \boxminus \mu_c) \boxplus \delta_{\sigma^2}$$

( $\boxminus$  = "the opposite of  $\boxplus$ ").

# Realization of the theorem for the problem at hand

Form the compound observation matrix

$$\begin{aligned}\hat{\mathbf{H}}_{1\dots L} &= \mathbf{H}_{1\dots L} + \frac{\sigma}{\sqrt{L}}\mathbf{X}_{1\dots L}, \text{ where} \\ \hat{\mathbf{H}}_{1\dots L} &= \frac{1}{\sqrt{L}} \left[ \hat{\mathbf{H}}_1, \hat{\mathbf{H}}_2, \dots, \hat{\mathbf{H}}_L \right], \\ \mathbf{H}_{1\dots L} &= \frac{1}{\sqrt{L}} \left[ \mathbf{H}, \mathbf{H}, \dots, \mathbf{H} \right], \\ \mathbf{X}_{1\dots L} &= \left[ \mathbf{X}_1, \mathbf{X}_2, \dots, \mathbf{X}_L \right].\end{aligned}$$

For the problem at hand, the theorem takes the form

$$\nu \frac{1}{m} \hat{\mathbf{H}}_{1\dots L} \hat{\mathbf{H}}_{1\dots L}^H \boxtimes \mu \frac{n}{mL} \approx \left( \nu \frac{1}{m} \mathbf{H}_{1\dots L} \mathbf{H}_{1\dots L}^H \boxtimes \mu \frac{n}{mL} \right) \boxplus \delta_{\sigma^2} \quad (5)$$

Since  $\frac{1}{m} \mathbf{H}_{1\dots L} \mathbf{H}_{1\dots L}^H = \frac{1}{m} \mathbf{H} \mathbf{H}^H$ , we can now estimate the moments of  $\frac{1}{m} \mathbf{H} \mathbf{H}^H$  from the moments of the observation matrix  $\frac{1}{m} \hat{\mathbf{H}}_{1\dots L} \hat{\mathbf{H}}_{1\dots L}^H$ , and thereby estimate the eigenvalues, and hence the channel capacity.

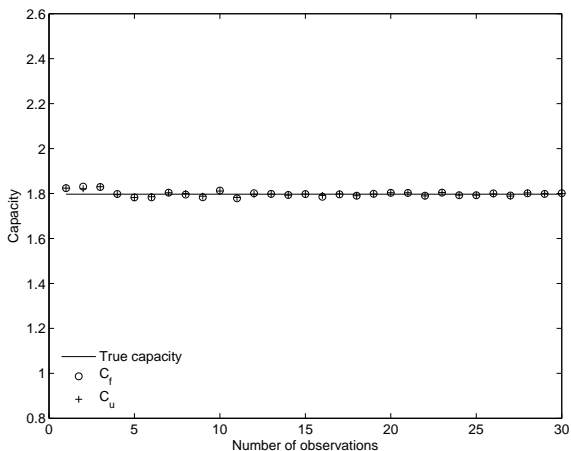


# Free probability based estimator for the moments of the channel matrix

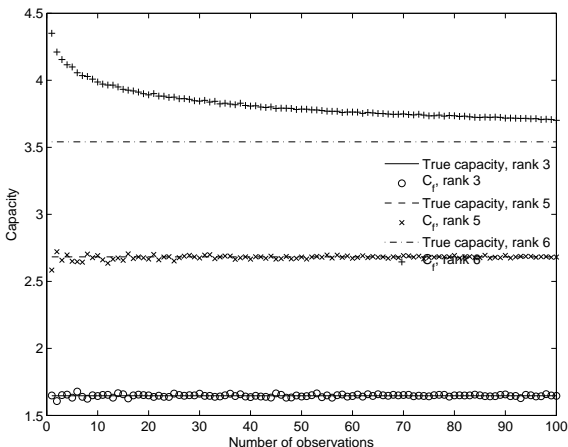
Can also be written in the following way for the first four moments:

$$\begin{aligned}
 \hat{h}_1 &= h_1 + \sigma^2 \\
 \hat{h}_2 &= h_2 + 2\sigma^2(1+c)h_1 + \sigma^4(1+c) \\
 \hat{h}_3 &= h_3 + 3\sigma^2(1+c)h_2 + 3\sigma^2ch_1^2 \\
 &\quad + 3\sigma^4(c^2 + 3c + 1)h_1 \\
 &\quad + \sigma^6(c^2 + 3c + 1) \\
 \hat{h}_4 &= h_4 + 4\sigma^2(1+c)h_3 + 8\sigma^2ch_2h_1 \\
 &\quad + \sigma^4(6c^2 + 16c + 6)h_2 \\
 &\quad + 14\sigma^4c(1+c)h_1^2 \\
 &\quad + 4\sigma^6(c^3 + 6c^2 + 6c + 1)h_1 \\
 &\quad + \sigma^8(c^3 + 6c^2 + 6c + 1),
 \end{aligned} \tag{6}$$

where  $\hat{h}_i$  are the moments of the observation matrix  $\frac{1}{m}\hat{\mathbf{H}}_{1\dots L}\hat{\mathbf{H}}_{1\dots L}^H$ ,  
 $h_i$  are the moments of  $\frac{1}{m}\mathbf{H}\mathbf{H}^H$ .



Comparison of  $C_f$  and  $C_u$  for various number of observations.  $\sigma^2 = 0.1$ ,  $n = 10$  receive antennas,  $m = 10$  transmit antennas. The rank of  $\mathbf{H}$  was 3.



$C_f$  for various number of observations. No phase off-set/phase drift.  $\sigma^2 = 0.1$ ,  $n = 10$  receive antennas,  $m = 10$  transmit antennas. The rank of  $\mathbf{H}$  was 3, 5 and 6.

# How would an algorithm for free convolution look?

## Definition

A family of unital  $*$ -subalgebras  $(A_i)_{i \in I}$  is called a free family if

$$\left\{ \begin{array}{l} a_j \in A_{i_j} \\ i_1 \neq i_2, i_2 \neq i_3, \dots, i_{n-1} \neq i_n \\ \phi(a_1) = \phi(a_2) = \dots = \phi(a_n) = 0 \end{array} \right\} \Rightarrow \phi(a_1 \cdots a_n) = 0. \quad (7)$$

A family of random variables  $a_i$  is called a free family if the algebras they generate form a free family.

- ▶ How do we implement this in terms of moments?
- ▶ From the previous result, we are basically interested in computing the moments of  $ab$ , when  $a, b$  are free, and  $b$  is free Poisson.

# Implementation of main result

The following formula can be used for incremental calculation of the moments of the measure  $\mu \boxtimes \mu_c$ , from the moments of the measure  $\mu$ :

$$[\text{coef}_m](cM_{\mu \boxtimes \mu_c}) = \sum_{k=1}^m [\text{coef}_k](cM_{\mu}) [\text{coef}_{m-k}] (1 + cM_{\mu \boxtimes \mu_c})^k. \quad (8)$$

Here,

- ▶  $M_{\mu}(z) = \mu_1 z + \mu_2 z^2 + \dots$ , where  $\mu_i$  are the moments of  $\mu$ .
- ▶  $\text{coef}_k$  means the coefficient of  $z^k$  in the polynomial.
- ▶ The power series coefficient can be computed through  $k$ -fold discrete (classical) convolution.
- ▶ (8) is proved by first proving that  $cM_{\mu \boxtimes \mu_c} = (cM_{\mu}) \boxtimes \text{Zeta}$ .

## Observation model 2

A more general observation model is:

$$\hat{\mathbf{H}}_i = \mathbf{D}_i^r \mathbf{H} \mathbf{D}_i^t + \sigma \mathbf{X}_i, \quad (9)$$

where  $\mathbf{D}_i^r$  and  $\mathbf{D}_i^t$  are  $n \times n$  and  $m \times m$  diagonal matrices which represent phase off-sets and phase drifts (impairments due to the antennas and not the channel) at the receiver and transmitter given respectively by

$$\begin{aligned} \mathbf{D}_i^r &= \text{diag}[e^{j\phi_1^i}, \dots, e^{j\phi_n^i}], \text{ and} \\ \mathbf{D}_i^t &= \text{diag}[e^{j\theta_1^i}, \dots, e^{j\theta_m^i}] \end{aligned}$$

where the phases  $\phi_j^i$  and  $\theta_j^i$  are random. We assume all phases independent and uniformly distributed.

# Problem when extending to phase off-set and phase drift

- ▶ In the compound observation matrix we now put

$$\mathbf{H}_{1\dots L} = \frac{1}{\sqrt{L}} [\mathbf{D}_i^r \mathbf{H} \mathbf{D}_i^t, \mathbf{D}_i^r \mathbf{H} \mathbf{D}_i^t, \dots, \mathbf{D}_i^r \mathbf{H} \mathbf{D}_i^t],$$

The moments of  $\frac{1}{m} \mathbf{H}_{1\dots L} \mathbf{H}_{1\dots L}^H$  are now in general different from the moments of  $\frac{1}{m} \mathbf{H} \mathbf{H}^H$ !

- ▶ In other words stacking the observations and using the free convolution framework does not give us what we want

A way to resolve this:

- ▶ Don't stack the observations at all.
- ▶ Perform convolution through exact formulas for the mixed moments of matrices and Gaussian matrices of lower order.
- ▶ Unbiased capacity estimator.

# Unbiased estimator for the moments of the channel matrix

Let  $\hat{h}_i$  be the first moments of the sample covariance matrix  $\frac{1}{m} \hat{\mathbf{H}}_i \hat{\mathbf{H}}_i^H$ . An unbiased estimator for the first moments  $h_i$  of  $\frac{1}{m} \mathbf{H} \mathbf{H}^H$  is given by

$$\begin{aligned}
 \hat{h}_1 &= h_1 + \sigma^2 \\
 \hat{h}_2 &= h_2 + 2\sigma^2(1+c)h_1 + \sigma^4(1+c) \\
 \hat{h}_3 &= h_3 + 3\sigma^2(1+c)h_2 + 3\sigma^2 c h_1^2 \\
 &\quad + 3\sigma^4 \left( c^2 + 3c + 1 + \frac{1}{m^2} \right) h_1 \\
 &\quad + \sigma^6 \left( c^2 + 3c + 1 + \frac{1}{m^2} \right) \\
 \hat{h}_4 &= h_4 + 4\sigma^2(1+c)h_3 + 8\sigma^2 c h_2 h_1 \\
 &\quad + \sigma^4 \left( 6c^2 + 16c + 6 + \frac{16}{m^2} \right) h_2 \\
 &\quad + 14\sigma^4 c (1+c) h_1^2 \\
 &\quad + 4\sigma^6 \left( c^3 + 6c^2 + 6c + 1 + \frac{5(c+1)}{m^2} \right) h_1 \\
 &\quad + \sigma^8 \left( c^3 + 6c^2 + 6c + 1 + \frac{5(c+1)}{m^2} \right),
 \end{aligned} \tag{10}$$



# Exact formulas for expectations of mixed moments of Gaussian and deterministic matrices

We have that

$$\begin{aligned}
 E [tr_n (\mathbf{W}_n)] &= m_1 + \sigma^2 \\
 E [tr_n (\mathbf{W}_n^2)] &= m_2 + 2\sigma^2(1+c)m_1 + \sigma^4(1+c) \\
 E [tr_n (\mathbf{W}_n^3)] &= m_3 + 3\sigma^2(1+c)m_2 + 3\sigma^2cm_1^2 \\
 &\quad + 3\sigma^4 \left( c^2 + 3c + 1 + \frac{1}{N^2} \right) m_1 \\
 &\quad + \sigma^6 \left( c^2 + 3c + 1 + \frac{1}{N^2} \right) \\
 E [tr_n (\mathbf{W}_n^4)] &= m_4 + 4\sigma^2(1+c)m_3 + 8\sigma^2cm_2m_1 \\
 &\quad + \sigma^4 \left( 6c^2 + 16c + 6 + \frac{16}{N^2} \right) m_2 \\
 &\quad + 14\sigma^4c(1+c)m_1^2 \\
 &\quad + 4\sigma^6 \left( c^3 + 6c^2 + 6c + 1 + \frac{5(c+1)}{N^2} \right) m_1 \\
 &\quad + \sigma^8 \left( c^3 + 6c^2 + 6c + 1 + \frac{5(c+1)}{N^2} \right), \tag{11}
 \end{aligned}$$

where  $m_j = tr_n \left( \left( \frac{1}{N} \mathbf{R}_n \mathbf{R}_n^H \right)^j \right)$ .

# Derivation of the limiting distribution for $\frac{1}{N}\mathbf{X}\mathbf{X}^H$

When  $x$  is standard complex Gaussian, we have that

$$E(|x|^{2p}) = p!.$$

A more general statement concerns a random matrix  $\frac{1}{N}\mathbf{X}\mathbf{X}^H$ , where  $\mathbf{X}$  is an  $n \times N$  random matrix with independent standard complex Gaussian entries. It is known [HT] that

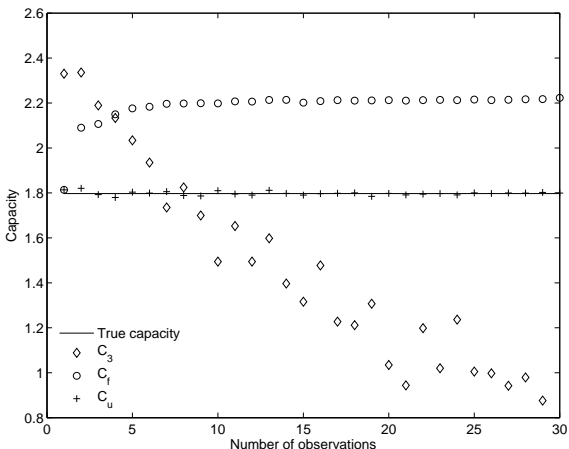
$$\tau_n \left( \left( \frac{1}{N}\mathbf{X}\mathbf{X}^H \right)^p \right) = \frac{1}{N^p n} \sum_{\pi \in S_p} N^{k(\hat{\pi})} n^{l(\hat{\pi})},$$

where  $\hat{\pi}$  is a permutation in  $S_{2p}$  constructed in a certain way from  $\pi$ , and  $k(\hat{\pi}), l(\hat{\pi})$  are functions taking values in  $\{0, 1, 2, \dots\}$ .

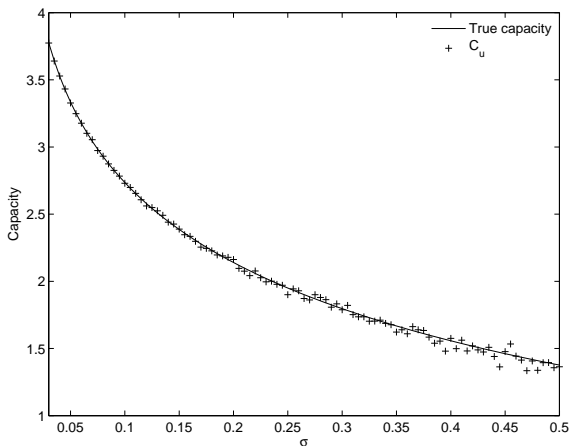
One can show that this equals

$$\tau_n \left( \left( \frac{1}{N} \mathbf{X} \mathbf{X}^H \right)^p \right) = \sum_{\hat{\pi} \in NC_{2p}} c^{l(\hat{\pi})-1} + \sum_k \frac{a_k}{N^{2k}}.$$

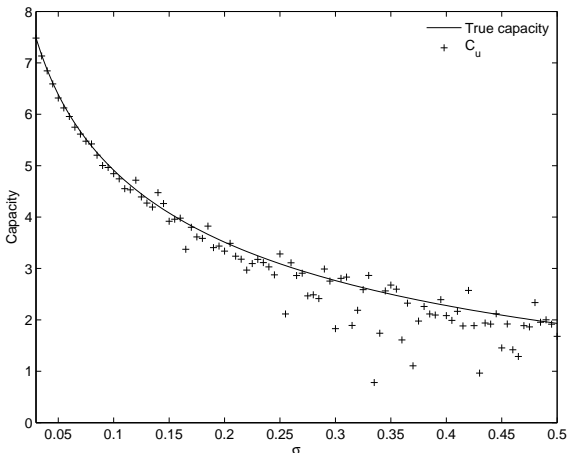
The convergence is "almost sure".



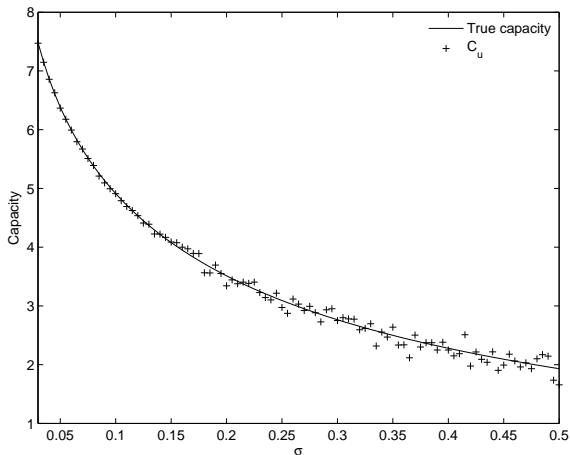
Comparison of capacity estimators which worked when no phase off-set/drift was present, for increasing number of observations. With phase drift and phase off-set.  $\sigma^2 = 0.1$ ,  $n = 10$  receive antennas,  $m = 10$  transmit antennas. The rank of  $\mathbf{H}$  was 3.



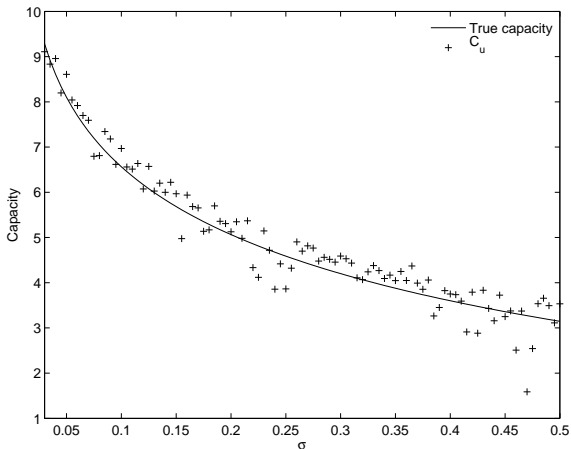
$C_u$  for  $L = 1$  observation,  $n = 10$  receive antennas,  $m = 10$  transmit antennas, with varying values of  $\sigma$ . With phase drift and phase off-set. The rank of  $\mathbf{H}$  was 3.



$C_u$  for  $L = 1$  observation,  $n = 4$  receive antennas,  $m = 4$  transmit antennas, with varying values of  $\sigma$ . With phase drift and phase off-set. The rank of  $\mathbf{H}$  was 3.

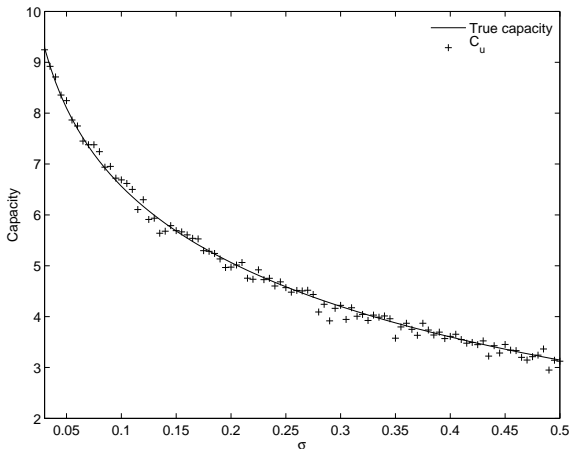


$C_u$  for  $L = 10$  observations,  $n = 4$  receive antennas,  $m = 4$  transmit antennas, with varying values of  $\sigma$ . With phase drift and phase off-set. The rank of  $\mathbf{H}$  was 3.



$C_u$  for  $L = 50$  observations,  $n = 4$  receive antennas,  $m = 4$  transmit antennas, with varying values of  $\sigma$ . With phase drift and phase off-set. The rank of  $\mathbf{H}$  was 4.





$C_u$  for  $L = 1600$  observations,  $n = 4$  receive antennas,  $m = 4$  transmit antennas, with varying values of  $\sigma$ . With phase drift and phase off-set. The rank of  $\mathbf{H}$  was 4.

# References

[HT]: "Random Matrices and K-theory for Exact  $C^*$ -algebras". U. Haagerup and S. Thorbjørnsen. [citeseer.ist.psu.edu/114210.html](http://citeseer.ist.psu.edu/114210.html). 1998.

This talk is available at

<http://heim.ifi.uio.no/~oyvindry/talks.shtml>.

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